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## Acoustic microsensors—the challenge behind microgravimetry

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**Abstract** Acoustic microsensors are commonly known as high-resolution mass-sensitive devices. This is a restricted view in many chemical and biosensor applications, especially in liquids. Sensitivity to non-gravimetric effects is a challenging feature of acoustic sensors. In this review we give an overview of recent developments in resonant sensors including micromachined devices and also list recent activity relating to the (bio)chemical interface of acoustic sensors. Major results from theoretical analysis of quartz crystal resonators, descriptive for all acoustic microsensors are summarized and non-gravimetric contributions to the sensor signal from viscoelasticity and interfacial effects are discussed. We finally conclude with some future perspectives.

**Keywords** Resonant microsensors · Acoustic sensors · Cantilever sensors

### Introduction

Acoustic microsensors are a versatile class of physical, chemical, and biological sensors known to be cost-effective, high-resolution, mass sensing devices. Their history began in 1959 when Sauerbrey published the essential relationship between the change of the resonant frequency of a quartz crystal and the mass added to its surface [1]. He consequently named the sensor the quartz crystal microbalance (QCM). Its use to monitor film thickness (growth) has been standard measurement technology for many decades in vacuum-deposition units. Other physical sensor applications with market relevance are temperature and pressure sensors, reviewed elsewhere [2]. The first application of quartz crystals as chemical sensors was reported by King [3]. The QCM has evolved to enable measurement in applications in analytical chemistry, electrochemistry, and biochemistry, because of its sensitivity at the transducer surface–analyte interface. It can

detect monolayers of small molecules, complex arrays of biopolymers and biomacromolecules, or whole cells.

The name microbalance implies that acoustic sensors measure mass or mass changes only. Indeed, in many applications acoustic sensors are used to convert a mass accumulated on the surface into a frequency shift. In chemical and, especially, biochemical applications, however, this basic understanding of the sensor principle can easily lead to misinterpretation of experimental results, especially when working in a liquid environment. It also hinders recognition by the experimenter of the outstanding capabilities of quartz crystal resonator sensors and other acoustic devices not available to other sensor principles. In a more general view acoustic sensors enable sensitive probing of changes within films attached to the transducer surface and at solid–solid and solid–liquid interfaces and are not restricted to mass changes [4]. Effects other than mass changes, so called non-gravimetric effects, can, moreover, be expected to contribute significantly to sensor response [5, 6]. Energy dissipation phenomena at its interface, a feature principally different from the mass effect, are just one obvious aspect.

Systematic application of non-gravimetric effects is a challenging task and requires a theoretical background both in chemistry or biochemistry, to enable prediction or estimation of specific property changes in the materials involved in the detection process, and in acoustics, to enable understanding of the path of signal transduction and estimation of changes in the (electrical) sensor signal. Manufacturing knowledge, including preparation technology, to optimize the signal-generation and transduction process of interest and to minimize the effect of unavoidable experimental variations, is also required.

In the text below we give a short review of recent developments of acoustic-wave-based sensors and the challenges behind microgravimetry. We start with a short overview of traditional and new acoustic sensors. It is not our intention to discuss chemical or biochemical aspects of acoustic sensors—we merely provide a list of very recent publications reporting improved chemical or biological sensitivity (for early papers, see references therein). The theoretical background of the sensor principle, based on

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acoustic wave propagation, will be summarized, but only as much as is necessary to enable understanding of the challenges of acoustic wave-based sensors behind microgravimetry. We also concentrate on quartz crystal resonators, because the results are representative for acoustic sensors. Finally we draw some perspectives of acoustic sensors.

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### Acoustic microsensors

The output signal of acoustic microsensors usually used is a frequency. Frequency is an analog value but can easily be converted into a digital signal, e.g. by counting the number of cycles in a defined period of time. In typical electrical circuitry the acoustic sensor is the frequency-determining element of an oscillator. Analysis of several solutions, ranging from oscillators to impedance analyzers, is given elsewhere, e.g. [7, 8]. Conversion of the mechanical resonant frequency into the electrical oscillation frequency does not cause loss of resolution. Transmission reliability of a frequency signal is very secure. Together with the high working frequency of acoustic devices it enables preparation of wireless sensors [9–11]. Frequency is insensitive to variations in the signal amplitude, drift, and bias. The vibration amplitude carries independent information and is therefore also used in modern acoustic sensor concepts. Because these advantages are common to all resonant sensors, it is not surprising that resonant sensors have attracted increasing interest in recent years.

Several acoustic devices have also been used as microsensors: surface acoustic wave (SAW) devices, surface transverse wave (STW) devices, flexural plate wave (FPW) devices, Love wave (LW) devices, and shear horizontal acoustic plate mode (SH-APM) devices are well known. For overviews of working principles, application features including (mass) sensitivity see, e.g., Refs. [12–15] and for graphic sketches, see, e.g. Ref. [16]. Some more specific devices not yet widely used are fiber acoustic wave (FAW) devices [17], tube acoustic wave devices [18, 19], and, most recently, a ball SAW device [20]. The concept of FAW devices is similar to FPW devices. A fiber can be as thin as a few micrometers, leading to very high sensitivity. Tube acoustic devices are very attractive from a packaging point of view, because the tube can act as liquid container; a measurement cell is not required. The sensitivity varies inversely with wall thickness. The advantage of ball SAW devices is a reduction of insertion losses, because of strong reduction of aperture. Film bulk acoustic resonator (FBAR) devices [21, 22] and micro-mechanical resonant structures, up to now mostly cantilevers, are the MEMS-based variant [23, 24]. Similar to the history of traditional acoustic sensors, improvements of the transducer element have been driven by different requirements, e.g. frequency normal [25–27] (quartz crystals), filter [28–31] (SAW, FBAR devices), and atomic-force microscopy [32–35] (cantilever).

Improvement of sensitivity is still a driving force in sensor development. Sticking again with the mass sensitivity argument, mass sensitivity of resonant sensors is basically governed by the ratio of mass change to the

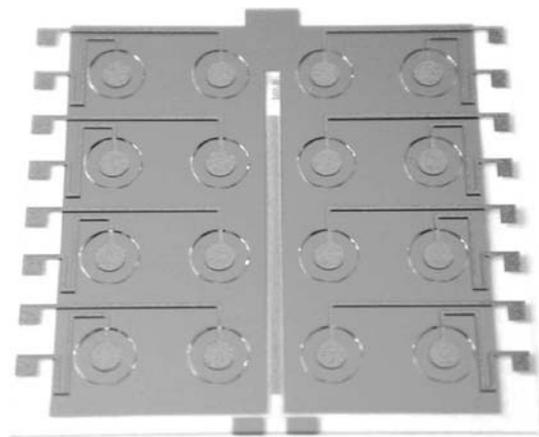
overall vibrating mass. As an approximation, the lighter the resonator the larger the relative mass increase, because of a bound mass. FPW devices have been a successful attempt to optimize mass sensitivity. In contrast to SAW devices (in which the vibrating mass is defined by the penetration depth of the acoustic wave and decreases with frequency) FPW devices can also be applied in liquids. Another attempt is quartz crystals with largely increased fundamental frequencies [36, 37]. These resonators have also been fabricated as arrays (Fig. 1) [38–40].

Another important development applies a technique known from non-destructive testing for acoustic wave generation. The so-called magnetic-acoustic-resonator sensor (MARS) uses a planar coil to generate an electromagnetic field at radio frequency [41, 42]. This field induces eddy currents in a metallic film deposited on the surface of a non-piezoelectric resonator. Lorentz forces appear in a strong static magnetic field and drive the resonator. They are capable of exciting several acoustic waves, including the shear mode, if the frequency of the applied current coincides with the respective resonant frequency of the resonator. Figure 2 shows the model used for FEM simulation. The key issue of this concept is a high quality factor ( $Q$ ) of the mechanical resonator. Therefore several materials have been tested. A  $Q$ -factor above  $10^5$  has recently been achieved with micromachined silicon membranes [43]. In a slightly different version the same principle has been applied to electrodeless quartz crystals [44, 45]. In this composition a D-field similar to that in the traditional configuration with electrodes arises, because of the specific properties of the permittivity tensor of AT-cut quartz [46]. One important advantageous feature of this concept is the capability of exciting harmonics up to the GHz range [47–49].

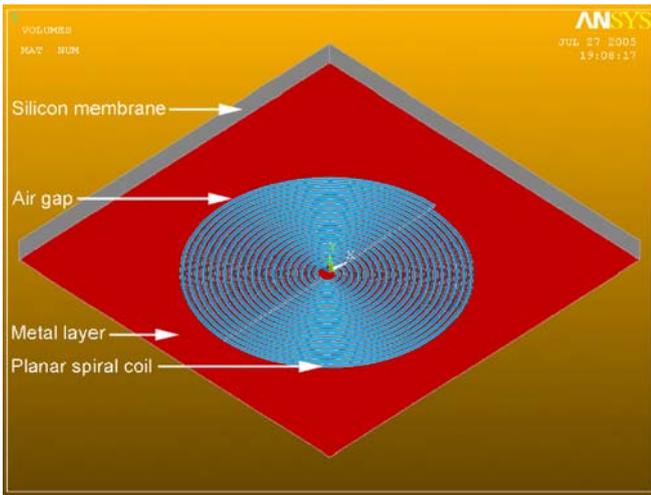
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### Micromachined devices

Silicon technology can lead to new possibilities—the capability to detect even smaller mass, the capability to



**Fig. 1** 4x4 quartz crystal sensor array. The membrane of each sensor element has been thinned by wet etching. The resonance frequency of each sensor could be elevated to frequencies up to 50 MHz [40]



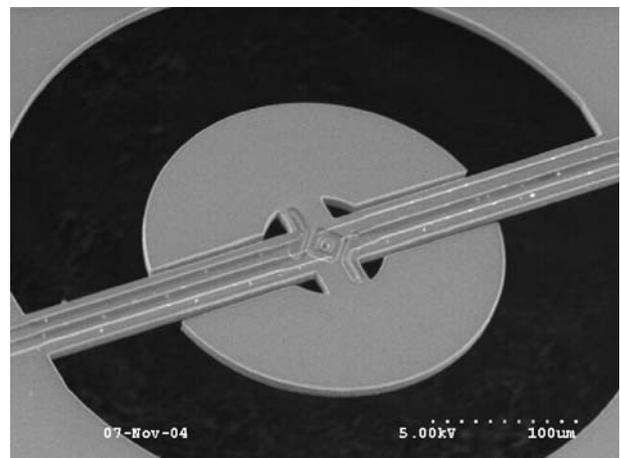
**Fig. 2** Basic model for FEM simulation. Eddy currents are caused by an RF current through the planar coil. Lorentz forces in the metal layer appear in a superposed static magnetic field, which is generated from a permanent magnet (not shown) [43]

fabricate arrays with a much larger number of elements per unit area, the capability of monolithically integrated electronic circuitry, and mass production at low cost. The available technology can also be adapted for sensor surface modification with specific coatings to create chemical selectivity. Nowadays commercial cantilevers are usually used. They are typically made of silicon, silicon nitride, or silicon dioxide. A great variety of dimensions and shapes is available. As sensors cantilevers can be used in the resonant or non-resonant mode. In the resonant mode mass deposited on to the cantilever reduces its resonant frequency. Assuming constant properties of the cantilever, frequency shift is proportional to mass change. The cantilever acts similar to the QCM. The proportional factor, and hence the mass sensitivity, depends on the force constant, which is a function of geometry and the apparent Young's modulus. In reality length and width can be well controlled during the etching process. The thickness of the beam is not usually known precisely enough. Although the Young's modulus of the cantilever also deviates from bulk values, accurate methods enable exact calibration of the sensor.

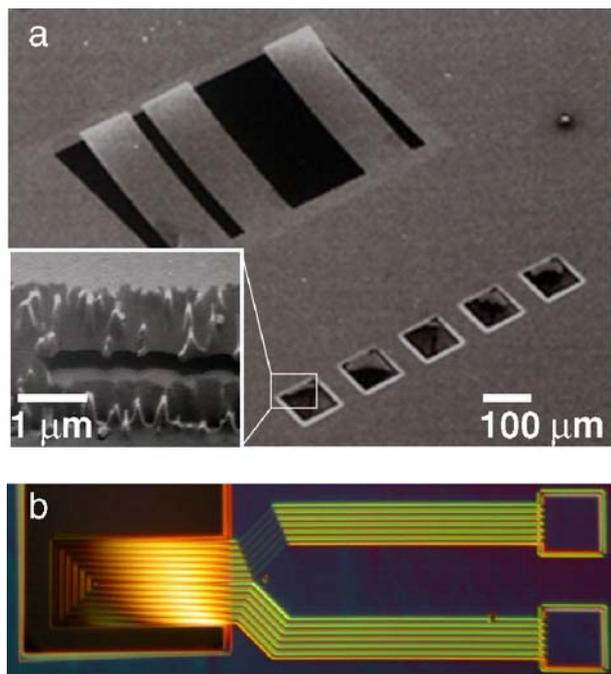
One example of a non-resonant application is the stress generated bending. Changes in surface stress can be the result of physical interaction, for example electrostatic forces between charged molecules on the surface, or of chemical nature, e.g. analyte absorption-induced swelling of a chemically sensitive coating during chemical sensing, conformational changes in a polymer film because of thermal treatment or radiation induced cross-linking, changes in surface energy, etc. In typical sensor arrangements this stress is not compensated at the opposite side of the cantilever and results in bending of the cantilever. This non-resonant application demonstrates the high sensitivity of those devices to factors other than mass. A review of cantilever applications in both principles of operation, including theoretical background, deflection detection techniques, and actuation methods is given elsewhere [50–52]; for very recent new activities see Refs. [53–62].

In liquid applications the resonant principle seems to be the more robust method whereas bending gives access to new fascinating applications based on structural changes in thin adsorbed layers. The resonant mode needs active driving of the cantilever and feedback for frequency measurement. Piezoelectric materials, for example PZT, are often used for that purpose; magnetic actuation [63] or pulsed laser heating [64] have also been applied. Cantilevered AT-cut quartz-crystal resonators have been fabricated by deep reactive ion etching [65]. One major challenging issue is an improved quality factor of the resonator.  $Q$ -factors of approximately  $10^3$  in the upper kHz frequency range in air enable a mass resolution in the picogram range. Magnetic actuation and a closed feedback loop can enhance the quality factor substantially [66]. A parametric resonance amplification has shown to provide a 1–2 orders of magnitude increased sensitivity compared with the simple harmonic resonance mode in air [67]. Another method for improved mass sensitivity is the use of higher modes. Because of the appearance of nodes additionally to the points of clamping, the resonant frequency change also depends on the position the mass is applied [68].

In a liquid environment, especially in biosensing applications, the out-of-plane vibration of the cantilever is strongly damped and results in an essentially reduced  $Q$ -factor of a few tens only. It can be enhanced by incorporating the cantilever in an amplifying feedback loop. With a monolithically integrated differential feedback circuit and a cantilever with electromagnetic actuation and piezoresistive readout a frequency stability of 3 Hz at a resonant frequency of about 200 kHz in water could be achieved [69]. Another approach avoids the out-of-plane vibration. A disc-shaped microstructure operates in a rotational in-plane mode with resonance frequencies between 300 and 700 kHz, Fig. 3. The open-loop  $Q$ -factor has been found to be as high as 5800 in air and 100 in water [70]. A very different approach avoids viscous damping by feeding the analyte through channels buried in the cantilever, Fig. 4 [71]. The cantilever still vibrates in air; the mass change occurs within the channels because of adsorption of the molecules of interest on the channel



**Fig. 3** SEM photograph of a high  $Q$  disk-shaped micromachined resonator [70] (reproduced with kind permission of J.H. Seo)



**Fig. 4** SEM micrographs of cantilever with buried microchannels [71] (reproduced with kind permission of S. Manalis)

walls. The sensor signal is actually based on density differences between the attached molecule and the buffer.

FBAR devices use a piezoelectric thin film for acoustic wave generation. They can have resonant frequencies in the GHz range. To keep this advantage the piezoelectric film is deposited on a micromachined silicon membrane. Another approach traps acoustic energy within a thin layer of a bulky silicon substrate, using several different layers with very different acoustic impedances, which act as acoustic mirror. This solution is especially interesting for sensing applications, because it solves the problem of fragility. The capability of this class of devices for chemical sensing has been reported elsewhere, e.g. Ref. [72]; for recent device developments and applications see Refs. [73–77].

Another group of acoustic sensors—ultrasonic sensors—should be mentioned here. This group of sensors shares some features with acoustic microsensors but there are also some remarkable differences. Similar to acoustic microsensors the acoustic wave is usually generated and detected with a piezoelectric device. In contrast, the acoustic wave travels through the bulk of the material of interest. Similarly, acoustic wave then carries information of both geometric and material properties of all materials along the acoustic path. Distance (level) or flow sensors and non-destructive testing are just two well known examples of a variety of applications of ultrasonic sensors.

Ultrasonic sensors have also proven their capabilities as chemical sensors, especially in process monitoring. These use the dependence of sound velocity, attenuation, and, nowadays, acoustic impedance on material properties relevant to process control, especially of complex liquid mixtures [78, 79]. Recently a MEMS version of ultrasonic sensors has been introduced, so-called capacitive micro-

machined ultrasonic transducers (cMUT) [80–82] and, recently, Refs. [83–86]. Instead of capacitive driven, piezoelectric excitation of vibration of a membrane at radio frequency is also possible [87]. Those transducers close the gap between acoustic microsensors and ultrasonic sensors and are very promising for microfluidic applications [88].

## Chemical interface

Acoustic sensors are uniquely sensitive to the mass of molecular species. This is a unique advantage but also requires precautions to avoid any unwanted interaction with other species. Cross sensitivity is therefore a major issue of the microbalance principle, whether with quartz crystals or other acoustic devices.

Acoustic sensors are inherently non-specific. The core of chemical analysis involving surfaces is a method for immobilization of the target molecule on the surface of the transducer. This issue is being addressed in various ways. King's approach to achieve chemical sensitivity and selectivity by modifying the surface of the quartz crystal with a chemically sensitive film is still standard technology. He started with materials used in gas chromatography as stationary phase. Meanwhile an extremely large variety of materials has been used for that purpose: numerous polymers [89–95], organometallic compounds [96–98], dendrimers [99–102], synthetic molecular recognition materials, including supramolecular substances [103], and molecular imprinted polymers [104–113] biomaterials [114–120] including DNA [121–128], proteins [129–140], cells [141] nanomaterials [142–146] and others.

The separation between transformation of a (bio)chemical signal into a physical signal in the film or on its surface and transformation of the physical signal into an electrical signal in the acoustic device is an advantage acoustic sensors share with other sensor principles. It enables use of the transducer device in many very different applications. The specification is mainly a question of selection or development of the appropriate sensitive material and the appropriate preparation technology and hence mainly a question of surface chemistry and application to complex (bio)molecular systems.

Other issues must be addressed to enable commercial competition with the success of other methods, especially surface plasmon resonance (SPR) instruments. A major challenge is a reproducible coating procedure without the need for expensive sensor calibration, irrespective of whether a self assembled monolayer, immobilization of a biomolecule, or a thick polymer film. It includes minimization of unspecific binding or any other loading of the sensor surface. The acoustic device must also be re-usable. Although inexpensive in principle, application of acoustic sensor in liquids, especially, requires devices with enhanced properties. Examples of cost-intensive special features of quartz-crystal sensors are a well polished smooth surface to prevent liquid trapping, specific electrode geometry to achieve a large grounded electrode, to prevent crosstalk to

the electrical properties of a contacting liquid, e.g. a buffer solution, and a contact stripe which goes round the edge to provide electrical contacts on one side of the crystal; gold as electrode material both to furnish a noble surface and also as the basis for stable and reliable surface functionalization is also required. Mechanical stability of the electrode is another major concern in terms of re-usability. The availability of commercially pre-activated acoustic sensors would greatly reduce the requirement for skilled operators, simplify measurement procedures and reduce the overall measurement time. As far as we are aware quartz crystals modified for specific applications are available from Q-Sense only [147].

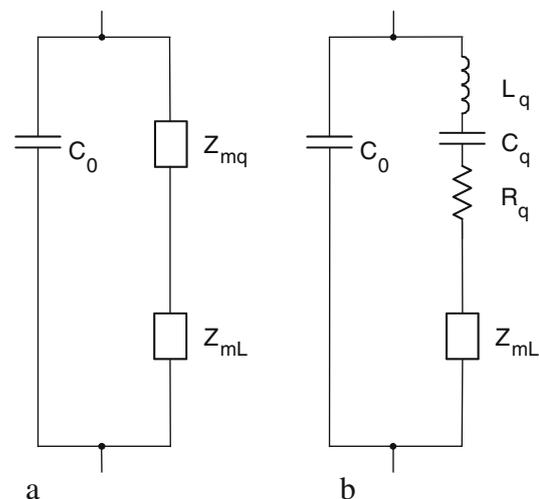
## Theoretical considerations

The theoretical background of electroacoustic devices is well developed. Several models are available, very basic physical models and equivalent circuit models [148–159]. In terms of description of sensor behavior they deliver very similar results. Acoustic microsensors are characterized by different prevailing acoustic modes. Quartz crystal resonators work with a thickness-shear wave, SAW sensors exploit a Rayleigh wave; both waves are transverse whereas the typical bending mode of cantilevers generates a longitudinal wave. It has, however, been shown that under typical experimental conditions acoustic microsensors transform the acoustic properties of the coating(s) and, if so, of a liquid, into equivalent electrical signals. When modeling wave propagation, one usually decomposes the wave vector into lateral components and a component normal to the main surface of the device. Important features common to all devices result from the vertical component. It obeys laws similar to those valid for quartz crystal resonators—hence the challenges behind the physics are also similar. The results derived for quartz crystal resonators cover the most important findings and are representative for acoustic microsensors. They are summarized below.

1. The basic principle of operation of a generic acoustic-wave sensor is a traveling wave in a confined structure which produces a set of standing waves whose frequency is determined jointly by the velocity of the traveling wave and the dimensions of the confining structure. The most basic method of resonator modeling consequently requires application of the theory of wave propagation taking into consideration the properties of the material and the geometric dimensions of the resonator.
2. A quartz crystal resonator (QCR) as used for chemical or biochemical sensing is a resonator composed of the piezoelectric crystal itself and at least one (non-piezoelectric) layer to achieve chemical sensitivity. In the transmission line model (TLM), one of the most successful tools for sensor analysis, a composite resonator, is represented as an arrangement of the respective number of transmission lines in series. Each transmission line has two acoustic ports and represents

one layer. The large aspect ratio between the lateral dimensions and the thickness of each layer enables one-dimensional treatment of wave propagation. The inherent assumption is that the material is homogeneous within one layer. The TLM also requires continuity of particle displacement and shear stress at the interface between all transmission line elements.

3. Acoustic properties of coating(s) and liquid are summarized in one value which is recognized by the piezoelectric transducer: the (effective surface) acoustic load (impedance),  $Z_L$  e.g. Ref. [160]. The acoustic load plays the central role for any chemical acoustic microsensor. It does not play any role if the acoustic load is generated by a simple mass (per area), as introduced by Sauerbrey, by a semi-infinite liquid, as first analyzed by Kanazawa, because of the viscoelastic properties of a single layer or from a multilayer arrangement. The acoustic load carries all acoustically relevant information, which is related to the chemical or biochemical interaction irrespective of whether a change in  $Z_L$  is generated by pure mass accumulation or if the interaction is accompanied by a change in material properties. Interfacial property changes between layers have also been found to produce significant changes of  $Z_L$ .
4. The TLM can be represented in an equivalent circuit consisting of two impedances in series parallel to  $C_0$ , Fig. 5.  $C_0$  is the only electrical element; the two impedances represent acoustic properties of the sensor. This branch is, therefore, called the motional arm.  $Z_{mq}$  represents the acoustic properties of the quartz crystal only. As is common for sensors,  $Z_{mq}$  is assumed to be constant.  $Z_{mL}$  is the electrical representation of the acoustic load,  $Z_L$ .
5. The plot of the electrical admittance vs. frequency of the circuitry (see, e.g., Ref. [161]) has several characteristic points which can be used for measurement, e.g. maximum or minimum of the admittance



**Fig. 5** Presentation of the transmission line model of a quartz crystal in a modified equivalent circuit (a) and the modified Butterworth–Van Dyke equivalent circuit (b).  $Z_{mq}$  represents the motional impedance of the quartz crystal,  $Z_{mL}$  is the electrical representation of  $Z_L$

magnitude or zero-phase crossing of the phase. The only straightforward resonant frequency is the maximum of the real part of the electrical admittance, also called in-phase admittance or conductance. The respective frequency,  $f_s$ , is the resonant frequency of the motional arm. Zero-phase oscillators require compensation of  $C_0$  to be applicable. Near resonance of the quartz crystal  $Z_{mq}$  can be presented as equivalent circuit elements  $C_q$ ,  $L_q$  and  $R_q$ , arranged as a serial resonant circuit. One obtains the modified Butterworth–Van Dyke equivalent circuit, Fig. 5b, e.g. Refs. [156, 162]. The complex element  $Z_{mL}$  can be separated into a frequency-independent resistance and inductance only in certain circumstances, e.g. for a Newtonian liquid.

6.  $Z_L$  is a complex value. The imaginary part of the acoustic load governs the changes in the serial resonant frequency [155], in most sensor applications the values are proportional:

$$\frac{\Delta f_s}{f_0} \propto -\text{Im}(Z_L) \quad (1)$$

7. The real part of the acoustic load describes acoustic energy loss within the layer(s). It can be represented by changes in the resistance of the serial branch of the equivalent circuit of the sensor e.g. Ref. [160], bandwidth [155] or a dissipation factor [163].

$$\Delta R \propto \text{Re}(Z_L) \quad (2)$$

8. Behling [164] introduced notation for the acoustic load which clearly depicts its character. He separated a mass factor  $M$  and an acoustic factor,  $V$

$$Z_L = jM \cdot V \quad (3)$$

where

$$M = \omega \rho h \quad (4a)$$

$$V = \frac{\tan \varphi}{\varphi} \quad (4b)$$

$\rho$  is the density,  $h$  is the thickness of the coating,  $\varphi$  is the phase shift the acoustic wave undergoes while traveling through the film and  $\omega = 2\pi f$  is the angular frequency,  $f$  the resonant frequency,  $j = \sqrt{-1}$ .

9. With the notation of Eq. 4a and 4b one can distinguish four regimes in which a QCR can work, Table 1.  $V=1$  represents the pure gravimetric response, the classical microbalance principle. In all other cases so-called non-gravimetric effects contribute to the sensor response.

**Table 1** Types of non-gravimetric regime

Case 1	$M \rightarrow M+dM$	$V=1$	$\Rightarrow$	Gravimetric sensor
Case 2	$M \rightarrow M+dM$	$V \approx \text{const} > 1$	$\Rightarrow$	Acoustically amplified gravimetric sensor
Case 3	$M \rightarrow M+dM$	$V \rightarrow V \pm dV$	$\Rightarrow$	Mass and material effect sensor
Case 4	$M \approx \text{const}$	$V \rightarrow V \pm dV$	$\Rightarrow$	Film properties sensor

They are called non-gravimetric because the phase shift and hence the acoustic factor contains material properties of the coating:

$$\varphi = \frac{M}{Z_c} \quad (5)$$

10.  $Z_c = \sqrt{\rho G}$  is the characteristic impedance of the coating material and should not be confused with the acoustic load.  $G$  is the shear modulus and a complex value (except for pure elastic materials, for which  $G$  is real, and pure viscous (Newtonian) liquids, for which  $G = j\omega\eta_l$  is imaginary;  $\eta_l$  is the viscosity).

11. For a complex shear modulus,  $G$ , the acoustic factor is also complex:  $V = V' + jV''$ , whereas  $M$  remains real. Consequently  $Z_L$  becomes:

$$Z_L = Z'_L + jZ''_L = -MV'' + jMV' \quad (6)$$

$V'$  can be understood as acoustic amplification of the mass effect.  $MV''$  is always a viscoelastic contribution to acoustic energy dissipation and hence to  $R$ ; a pure mass does not change  $R$ .

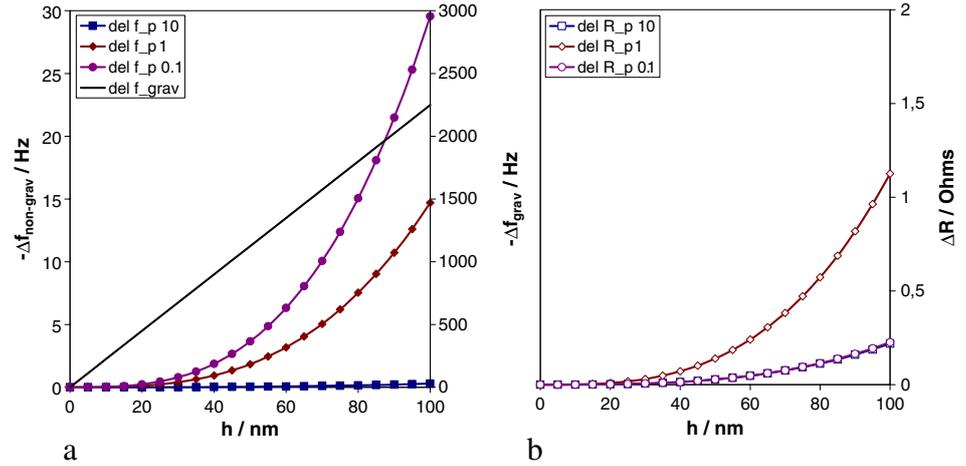
12. Behling's notation can also be applied to a two-layer arrangement, e.g. a film covered with a liquid:

$$Z_L = j(M_1 V_1 + M_2 V_2) V_M \quad (7)$$

where  $V_M = (1 - M_1 V_1 M_2 V_2 / Z_{c1}^2)^{-1}$  is the mixed acoustic factor [165]. The additive character of the contributions of both layers is given only for  $V_M = 1$ , i.e.,  $Z_{c1} \gg M_1 V_1 M_2 V_2$  which usually requires a large  $G$  for the first film.

The gravimetric regime, Table 1 case 1, is obviously the simplest way of exploiting the capabilities of acoustic microsensors. The proof that indeed  $V=1$  is not that simple, as stated below. The shift in frequency because of absorption or adsorption of species can be directly related to the mass accumulated—in principle even without calibration (device characteristics must be known). Under real conditions the electrical and acoustic properties of the transducer in the actual measurement arrangement must be determined. This is especially true for liquid applications, in which surface roughness (liquid trapping), sensor sealing (stress genera-

**Fig. 6** Gravimetric frequency shift (**a**, right axis) and signature of viscoelastic contributions to the sensor response (**a**, left axis, and **b**) of a single rubbery film ( $G'=1$  MPa) and a loss factor  $\tan \delta=10, 1$ , and  $0.1$ , respectively (see legend). The resonant frequency is  $10$  MHz



tion, spurious modes), and the properties of the sensor electronics (phase stability, capacitance compensation) introduce unwanted variances in the sensor transfer function. Unfortunately, many authors stick to frequency changes and consequently do not report on calibration issues. It is not surprising that absolute values differ for similar experiments from one laboratory to another. One can expect helpful information from a ring experiment to extend the large success of acoustic microsensors in laboratory applications to the sensor market.

The second issue is sensor sensitivity which is limited in the gravimetric regime and cannot compete with optical methods such as SPR. Although MEMS-based sensors can significantly improve sensor sensitivity, the signal-to-noise-ratio limitation because of a low  $Q$ -factor persists.

## Challenges from non-gravimetric effects

### Effect of single-film viscoelasticity

Viscoelastic materials are characterized by a complex shear modulus.  $G$  of macromolecular materials (including biomolecules) may vary between  $10^4$  Pa and  $10^9$  Pa. Rigid or glassy materials have a storage modulus,  $G'$ , larger than  $10^8$  Pa and a loss modulus,  $G''$ , which is an order of magnitude smaller. Rubbery materials have a modulus of approximately  $10^6$  Pa, or smaller. A large variety of material property combinations is conceivable. For analysis of acoustic consequences variation of the loss factor  $0.1 \leq \tan \delta = G''/G' \leq 10$  is sufficient.

In chemical and biochemical sensing the chemically sensitive film is usually thin and the phase shift the acoustic wave undergoes is small ( $\varphi$  usually smaller than  $0.2$ ). In these circumstances the approximation  $\varphi = \varphi + \varphi^3/3$  can be applied to Eqs. 1, 2, 3, 4a and 4b. It results in [155, 163, 164]:

$$Z_L = jM \left( 1 + \frac{1}{3} \frac{M^2}{Z_{c1}^2} \right) \quad (8)$$

$$\Delta f \propto MV' = M \left( 1 + \frac{1}{3} \frac{J'}{\rho} M^2 \right) \quad (9a)$$

$$\Delta R \propto -MV' = M \left( \frac{1}{3} \frac{J''}{\rho} M^2 \right) \quad (9b)$$

introducing the compliance with  $J' = G'/(G'^2 + G''^2)$  and  $J'' = G''/(G'^2 + G''^2)$ . The viscoelastic contribution to the frequency shift, i.e. the second term in the brackets in Eq. 9a, depends on  $M^2$ , i.e., it depends on the film thickness and, consequently, vanishes for very thin films. In the following treatment we apply the TLM to a 10-MHz AT-cut quartz crystal with an electrode diameter of 6 mm and material properties typical of those of macromolecular substances and a film thickness range also typical of chemical and biosensors. Therefore this and the following diagrams should be understood as a signature of the sensor response [165]. Figure 6a shows, on the right ordinate, the gravimetric response to a single rubbery film facing air vs. the thickness of the layer and, on the left ordinate, the viscoelastic contribution to the sensor response. Compared with the gravimetric response the non-gravimetric contribution to the frequency shift is very small for the conditions selected. Figure 6b shows the resistance increase because of viscoelasticity. It is almost negligible.

### Effect of the viscoelasticity of a single film facing a liquid

We replicate the same analysis but now the viscoelastic film faces a liquid, here water. Repeating the above

mathematical treatment, also applying the approximation:  $1/(1-x) \approx 1+x$  results in:

$$\Delta f \propto M \left( 1 - 2L^2 \frac{J'''}{\rho} \right) \quad (10a)$$

$$\Delta R \propto M \left( 2L^2 \frac{J''}{\rho} \right) \quad (10b)$$

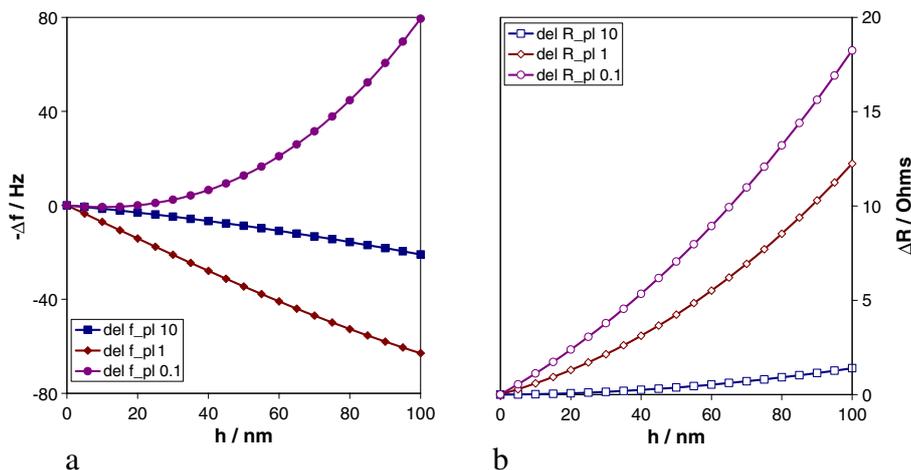
$L$  replaces  $(\omega\rho_i\eta_i/2)^{1/2}$  and summarizes liquid properties. Note the negative sign in Eq. 10a as opposed to Eq. 9a. Kasemo called this effect “missing mass” [169]. Furthermore,  $V$  does not depend on the mass factor. Hence this term appears even for very thin films. The poor quality of the above approximations with increasing film thickness can be improved with an additive term within the brackets linearly dependent on  $M$ :  $\Delta f : L \frac{J' - J'''}{\rho} M$ ,  $\Delta R : 2L \frac{J' + J'''}{\rho} M$ . The sign of the extension term of Eq. 10a depends on the loss factor. It vanishes with  $\tan \delta = 0$  and is positive for materials with  $\tan \delta < 1$ , thereby compensating the missing mass effect with increasing film thickness. The resistance always increases, in contrast with the behavior of the frequency shift.

Figure 7 reflects these properties. The slope of the viscoelastic contribution to the frequency shift is negative in two cases. This contribution does not vanish even for very thin films. The absolute values are significantly larger, although still small compared with the gravimetric frequency shift (Fig. 6a). The resistance increase is more pronounced. Finally, the contribution to the frequency shift of the elastic film in water has a negative slope for very small  $h$  but it becomes positive with increasing  $h$  and the “extra mass” value exceeds the value in air for larger film thickness.

### Interfacial effects

Interfacial effects related to material properties near the surface which are different from bulk properties can be

**Fig. 7** Signature of the viscoelastic contribution of the same film as in Fig. 6, but the film now faces water. The gravimetric contribution and the contribution of water to the sensor response have been subtracted



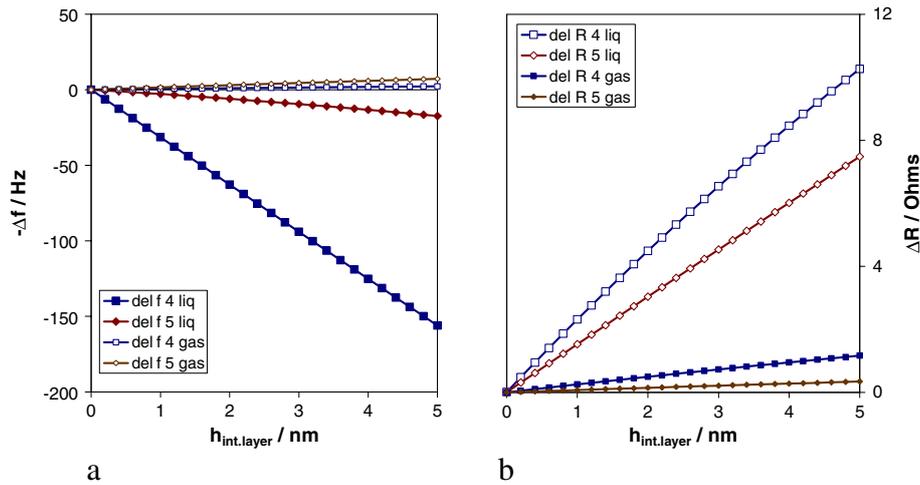
treated by introducing sub-layers. Interfacial slip [170–179], a violation of the continuity assumption, can also be modeled in this manner. The general acoustic understanding of interfacial phenomena is the appearance of a phase shift,  $\varphi_{\text{int}}$  and a characteristic “interfacial” impedance [174]. Figure 8 shows the signature of a (hypothetical) water-like interfacial layer with two different elastic components. Viscous and elastic behavior may arise, for example, if the interfacial layer consists of molecularly distributed water molecules and linker molecules. Other interfacial arrangements are conceivable [180–184]. The effect of an interfacial layer on the frequency shift can be remarkable in a liquid environment and does not depend very much on the shear modulus of the film above. A soft film with a shear modulus of  $G' = G'' = 1$  MPa would respond almost similarly to the glassy film, except that the viscoelastic contribution shown in Fig. 7a would appear as an offset. For comparison, the appearance of an interfacial layer in air is almost negligible.

### Changes in viscoelasticity and interfacial properties

Deviations from the gravimetric response are measurable and perhaps useful for viscoelastic amplification of the mass effect—Table 1 case 2. Arrangements for this purpose have been analyzed elsewhere [165]. Sufficient reproducibility is difficult to achieve, because of unavoidable experimental variations and uncertainties in film thickness or surface coverage. In particular,  $R$  (or any other value reflecting damping) has other sources than viscosity and tends to vary from one sensor to the next.

A more interesting aspect of a chemical or biochemical experiment is, however, the response of the sensor after interaction of a specific molecule with an already existing film [168]. Now  $dZ_L$  is of interest:

$$dZ_L = \frac{\partial Z_L}{\partial M} dM + \frac{\partial Z_L}{\partial V} dV \quad (11a)$$



**Fig. 8** Signature of an interfacial layer at a solid–solid interface (quartz crystal macromolecular film) with  $G'=10^4$  Pa or  $10^5$  Pa, exponent see legend, and  $G''=10^{4.8}$  Pa, equivalent to a viscosity of 1 cP at the resonant frequency of 10 MHz. The film is rigid and has a shear storage modulus  $G'=1$  GPa and a shear loss modulus  $G''=$

0.1 GPa. The film faces air (gas) or is covered with water (liq). The overall thickness of the film and interfacial layer has been kept constant at 50 nm. The gravimetric contribution of the primary film (50 nm) and the contribution of water to the sensor response have been subtracted

with

$$dV \approx \frac{\partial V}{\partial M} dM + \frac{\partial V}{\partial G} dG \quad (11b)$$

The second term in Eq. 11a cannot be neglected, because of changes in the shear modulus, whereas density changes do not play a recognizable role and are therefore omitted in Eq. 11b. Eq. 11a and 11b result in a viscoelastic contribution to the frequency shift of a single film (Eq. 9a) which is already three times larger even with  $dG=0$ .

Table 1 case 2 is difficult to distinguish from case 1, at least for gas sensors. Within experimental error one obtains almost linear dependence of the frequency shift on analyte concentration and the resistance increase is still not obvious. If viscoelastic contributions to the frequency shift are not noticed, calculated primary values, for example accumulated mass, or secondary values, for example partition coefficients, are too large if Sauerbrey's equation has been used.

Table 1 case 3 leads to a means of characterizing the materials of thin polymer films or, remembering the small penetration depth of shear waves in liquids, the interfacial properties of liquids, for example in lubrication or nanotribology [185–195].

Table 1 case 4 is most challenging, because biochemical problems are often related to small molecules, e.g. an enzyme interacting with a large protein. This interaction can change the conformation of the protein, which is accompanied by a modulus change, as is required for case 4. The extreme sensitivity of acoustic microsensors is because the shear modulus can easily vary by an order of magnitude or more. Changes of refractive index are, in contrast, in the percentage range. This is an outstanding trait which makes acoustic sensors extremely versatile in all applications in which the stiffness of thin layer(s) may change. The huge

potential of case 4 is about to be discovered for biosensors [196–199]. In case 4 the sensor acts almost purely non-gravimetrically. In this case, with an almost constant mass factor,  $M$ , both  $\Delta f$  and  $\Delta R$  will change because of changes in  $G'$ ,  $G''$ , and/or  $\tan \delta$ .

A similar sensor signal may have its origin in interfacial property changes. The non-gravimetric contributions of an interfacial layer may exceed viscoelastic contributions. Although one would not expect large changes in the thickness of an interfacial layer, large frequency and resistance changes can be expected, because of changes in the elastic properties of this layer, as is obvious when moving between the curves in Fig. 8. Note, that interfacial contributions can be constructive or opposite to viscoelastic contributions. It depends very much on the actual viscoelastic properties of both the interfacial layer and the thin film and how they change. These interfacial effects would, however, not appear when using cantilever-like sensors. Longitudinal waves do not probe apparent shear elasticity of an interfacial layer as analyzed above. Incompressibility of the interfacial liquid would dominate.

Prediction of sensor signals as illustrated by Figs. 6–8 is quite helpful; analysis of experimental results is much more involved. As discussed above, the sensor signal just depends on two quantities—the real and imaginary part of the acoustic load. The number of unknowns, even in the simplest case of a single film in air, is four and it increases the more complicated the model becomes. Care must be taken if literature values, usually bulk material properties, are used. The values may differ for thin films or at high probing frequencies. One way to expand the number of independent experimental data is to perform measurements also at overtones of the resonator [200–203]; the frequency dependence of  $G$  must also be taken into account, however [204].

## Other sensor signal sources

Nuisances are encountered in real experimental arrangements; these nuisances cannot be treated with simple theoretical models. Some work has been published dealing with surface roughness [205–208], stress [209], non-uniformity [210], or compressional waves [211–216]. A major concern is that these factors tend to be affected by the experiment and, therefore, contribute to the measurement signal in an uncontrolled manner. Generalization of acoustic characteristics enables partial treatment within the TLM [217].

Acoustoelectric effects, also, have not yet been considered. Electrode geometry, electrode polarity, fringing fields, and liquid conductivity or liquid permittivity affect the performance of quartz crystal resonator sensors operating in electrolytes and solutions. Electroacoustic effects result from acoustic wave propagation on SAW-like sensors. Here the acoustic wave propagates along the surface of a piezoelectric crystal and interacts also via the associated electrical fields with the overlayer and probes the electrical properties of the material, e.g. the permittivity of the film or the adjacent liquid. For example, the conductivity of the film gives rise to an additional decrease in acoustic wave velocity and to a peak in attenuation [218]. To take advantage of acoustoelectric effects with quartz crystals specific electrode geometry is required [219–223]. A specific version is lateral field excitation [224–227].

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## Perspectives

There is, understandably, continuing interest in gravimetric acoustic sensors. Their principle of detection is the simplest. Molecular weight is surely a fundamental property of species; therefore, to name two advantages, it does not require any charge transfer or specific labeling. Many acoustic sensors work well in viscous liquids and optically opaque media. The technique is quite easy to use and the equipment required is inexpensive. The sensitivity, however, is limited and cannot compete with that of optical methods, e.g. SPR, and so the gravimetric acoustic principle has a better chance of being applied in simple sensors than in scientific instrumentation. Improvement of sensitivity by use of additional sensor-system elements has been achieved with pre-concentrators and selectivity has been improved by use of chromatography. The range of sensor applications is broad; more applications can be expected for liquid systems, e.g. detection of water pollutants and health monitoring, for which competition from other sensor principles is smaller. Liquid applications require further improvements in sensor-system design to enable their use by unskilled persons also. Contactless magnetic excitation is a promising technique for simplifying handling and overcoming the restriction to piezoelectric materials for resonators but needs further improvement of energy-transfer efficiency. Micromachined sensors, including the development of sensor arrays adapted to robots

used in high-throughput technology, overcome the lack of integration potential of classical acoustic devices. The development of these is just beginning. One can expect the necessary improvements, especially in the  $Q$ -factor, from vibration modes not common for (atomic) force sensors. Sensitivity to mechanical damage must also be addressed.

Sensor applications also require improved sensor-to-sensor reproducibility, to reduce calibration effort. In practice, frequency measurement before use should be sufficient. Even a reference measurement with a reference substance makes measurement more involved and increases cost; this is acceptable in special cases only. Because reproducibility depends mainly on the reproducibility of the coating procedure, commercially pre-functionalized sensors may lead to an increase in market confidence in acoustic microsensors.

Non-gravimetric acoustic sensors are a much greater challenge. These acoustic sensors can provide measurement capabilities not available with other methods. The outstanding sensitivity of acoustic wave-based sensors to materials and interfacial properties is still far from being systematically exploited. One reason is the lack of relevant information about those properties. The development of non-gravimetric measurement techniques can positively affect research areas such as polymer characterization or micro/nanotribology and can, especially, boost measurement capabilities in biochemistry or pharmaceutical and biomaterials research. Growing confluence of previously unrelated disciplines is required. Miniaturization on micro and nanoscales, array technology, and improvements in signal-transduction mechanisms combined with information processing has created a rapidly developing field of measurement technology. The theory of acoustic wave propagation can be used to predict sensor response to material, interfacial, and other properties, further effects, for example mass point contact, are about to be described. It is now necessary to develop better models of the (bio) systems themselves, probably on the level of molecular dynamics. Improvements in preparation technology enable access to systematic variation of acoustically relevant properties. Finally, other measurement technology [228–235] must be simultaneously applied to reduce the number of unknowns of the systems under investigation.

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